



### A 187 dB CIFF Delta-Sigma Modulator

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#### ABSTRACT

This paper presents a single loop sixth-order delta-sigma modulator topology of cascade of integrator with multiple feedforward (CIFF) having multi-bit quantizer. The quantizer is 3bit to enhance the levels of quantizer. The modulator noise transfer function (NTF) and signal transfer function (STF) for optimized. The out-of-band gain (OBG) of 3 selected considering the stability of loop. Due to multi-bit quantizer, the modulator can achieve higher performance. The NTF zeroes optimized for maximum quantization noise suppression. A full-scale signal 0.55-V optimized for maximum flat response as compared to cascade of integrator with multiple feedback (CIFB). The key advantage of CIFF topology is to use only one digital-to-analog converter (DAC), it simplifies the feedback DAC implementation. The NTF zeroes are spread on the DC on the unit circle to maximize the performance of the modulator. The complete modulator has six integrator in the loop filter with single feedback DAC. The operational amplifier inside the integrator in the loop filter optimized. The complete modulator simulation shows it can achieve signal-to-noise-ratio (SNR) of 187 dB with oversampling ratio (OSR) of 128.

**Keywords:** Signal Transfer Function, Noise Transfer function, Unit Circle, Quantizer, operational amplifier

#### 1. INTRODUCTION

The scaling down of CMOS technology continuously pushes digital systems to have higher density, higher speed, and lower cost. The technology scaling is generally associated with a decrease in supply voltage, which reduces the signal swing and circuit dynamic range [1]. This trend brings great difficulties to the design of a high-resolution analog-todigital converter (ADC). Meanwhile, low power consumption becomes more critical in modern integrated circuit design to meet the increasing demands of the battery-powered portable device [2]. A fourth-order multi-bit cascade-ofmultiple-feedback (CIFB) delta-sigma modulator ADC for Low bandwidth applications. The proposed modulator topology can achieve 30-bit resolution for very small bandwidth application. Recently delta-sigma modulator the gaining popularity for higher signal applications. bandwidth While the continuous-time delta-sigma modulator have key advantages does not require anti-aliasing filter and requires much lower gain-bandwidth. Discreate-time delta-sigma modulator have switchedcapacitor implementation with components of capacitor, switches, and operational amplifier as an integrator. The continuous-time (CT) implementation uses resistor and capacitor with operational amplifier as integrator. It is well known that Nyquist pipeline ADC require accurate inter-stage gain, that determine high-gain wideband residue amplifier and calibration technology, leads to complexity and power efficiency employing noise shaping and





oversampling. However, the requirement of oversampling ratios (OSRs), that is typically small [3-6]. The delta-sigma modulator more than 50 MHz getting popularity due to increased demands in system. discrete-time portable The implementation quite popular due to high accuracy but smaller bandwidth. The demand for higher bandwidth rising with increased resolution. The challenges due to smaller OSRs due to the reason of process limited clock rate. To reach sufficient higher dynamic range, higher order noise shaping needs to be implemented by increasing the noise function order, transfer that is conventionally performed by loop filter cascade and generally equal or greater than three or more required. At the same time increased order of the loop filter causes stability issues [7]. A wideband continuous-time third order with 4-bit quantizer delta-sigma modulator designed for signal band of 100MHz can achieve dynamic range of 80-dB in 40 nm CMOS Technology at supply voltage of 1.2 V with power consumption of 69.7 mW. The modulator three-stage uses feedforward amplifier with miller compensation. The first integrator has unity gain bandwidth of 3.6 GHz and phase margin of 57 degree including all loading effect with power consumption of 10.5mW from power supply of 1.2V. The second and third integrator adopts the similar structure of the amplifier with scaled bias currents, that results in achieving unity gain bandwidth of 4.7 GHz and 3.3 GHz respectively. The phase margin for second and third integrator is 58 degree and 57 degrees and consuming power 4.3mW 17.3mW of and respectively. The power breakdown shows that loop filter consumes more than 46 % of the powers with total modulator power of 69. 7mW. While the 4-bit guantizer implemented as 4-bit flash ADC with each cell uses preamplifier due to smaller signal swing in front of quantizer. The digital-toanalog converter (DAC) is implemented using current steering unit element. Finally, the modulator can achieve signato-noise-plus-distortion ratio (SNDR) of 76dB and DR of 84 dB for signal bandwidth of 100 MHz for 40 nm CMOS Technology [8]. A CT modulator with third order loopfilter with preliminary sampling and quantization scheme in backend sub-ranging multi-bit quantizer can achieve SNDR of 72 dB for signal bandwidth of 100 MHz. The cascade of integrator with multiple feedforward topology used to take advantage of single DAC. While segmented non-return to zero (NRZ) DAC used for the feedback. The excess loop delay [ELD] compensation implemented around the second integrator. The modulator total power consumption is 16.3mW, while the loop filter integrators consume 9.4mW, DAC consumes 3.5mW, quantizer 1.4mW and CLK and digital consumes 2mW respectively in 28 nm CMOS Technology sampling at 2GS/s. The modulator can achieve SNDR of 72.6dB, SNR 73.2 dB and dynamic range of 76.3 dB respectively for signal bandwidth of 100MHz in 28nm CMOS Technology [9]. Another high bandwidth CT delta-sigma modulator with signal bandwidth of 125MHz VCO proposed based on based integrators. The modulator uses VCO based quantizer and segmented phasedomain ELD compensation. The loopfilter is active RC integrator based, while modulator topology is cascaded of integrator with multiple feedforward (CIFF). The first integrator operational amplifier unity gain frequency is four times to sampling frequency. The second integrator operational amplifier unity gain frequency is three times to the modulator sampling frequency. While the





third RC integrator operational amplifier unity gain frequency will be three times to sampling frequency. Finally, the third order modulator with VCO based quantizer can achieve SNDR of 71.9 dB and dynamic range of 74.8dB for signal band of 125MHz with power consumption of 54mW sampling frequency of 2.15GHz with OSR of 8.6 in 16nm CMOS Technology [10]. A CT Multi-stage Noise Shaping (MASH) modulator designed for signal band of 465MHz in 28 nm CMOS Technology. A 1-2 MASH topology is adopted to achieve aggressive noise shaping with higher stability at much lower OSR of 8.6. The first stage is a first order modulator to reduce the power amplifier for given thermal noise requirement. The first stage consists of an active RC integrator, quantizer with 17-level and current steering DAC and capacitive DAC. The second stage consists of an active-RC resonator, flash ADC of 17-levels, current steering DAC and capacitive DAC used to provide fast direct feedback. The second stage uses feedback topology to minimize STF peaking and the input full-scale of second stage is scaled down to provide an interstage gain of six to minimize the overall quantization noise floor while preventing the residue of the first stage from saturating the second stage [11].

This paper a single loop sixth-order deltasigma modulator topology of cascade of integrator with multiple feedforward (CIFF) having multi-bit quantizer. The quantizer is 3-bit to enhance the levels of quantizer. The modulator noise transfer function (NTF) and signal transfer function (STF) for optimized. The out-ofband gain (OBG) of 3 selected considering the stability of loop. Due to multi-bit quantizer, the modulator can achieve higher performance. The NTF zeroes optimized for maximum quantization noise suppression. A full-scale signal 0.55-V optimized for higher order loop stability. Due to CIFF topology, peaking in the STF is optimized for maximum flat response as compared to cascade of integrator with multiple feedback (CIFB). The key advantage of CIFF topology is to use only one digital-to-analog converter (DAC), it simplifies the feedback DAC implementation. The NTF zeroes are spread on the DC on the unit circle to maximize the performance of the modulator. The complete modulator has six integrators in the loop filter with single feedback DAC. The operational amplifier inside the integrator in the loop filter optimized. The complete modulator simulation shows it can achieve signal-tonoise-ratio (SNR) of 187 dB with oversampling ratio (OSR) of 128.

After the introduction, the second discuss the design section of the modulator design with CIFF topology, while the third section describes the modeling and simulation of the modulator and explain with reference to the non-idealities for the fourth-order multi-bit quantizer for discrete-time implementation. Finally, the section four concludes the paper.

# 2. MODULATOR DESIGN

A fourth-order multi-bit modulator modeled using Delta-Sigma Toolbox [12].

## Table I : CIFB Cofficients

Parameters	Values
a1	0.292
a <sub>2</sub>	1.421
<b>a</b> <sub>3</sub>	2.744
a4	2.553
b <sub>1</sub>	0.292
<b>g</b> <sub>1</sub>	0.0
Ø <sub>2</sub>	0 000411





The cascade of integrator with multiple feedforward(CIFF) investigated with higher out-of-band-gain (OBG). Also, to boost the performance of the modulator, the NTF zero optimization techniques is employed. The CIFB topology can achieve SNR of 182 dB with oversampling ration (OSR) of 131. The CIFF topology allows higher stability with multiple to feedback. The NTF zero optimization causes more in-band guantization noise shaped to out-of-band. Without NTF zero



optimization technique the modulator topologies can achieve SNR of 175 dB with OSR of 131 and OBG of 4. The CIFB modulator coefficient obtained from the toolbox shown in Table-I. This coefficient represents the ratio of capacitors at the switched-capacitor implementation, for the discrete-time implementation of the modulator. The signal-transfer function (STF) and noise transfer function (NTF) of the modulator is shown in Figure 1. The STF shows no peaking with smooth response as low pass filter response. While the NTF of the modulator shown in Figure 1 is a high pass response. While the Figure 2 shows the output stages of each integrator, due to CIFB topology the swing inside the loop filter is large and it demands operational amplifier with high open loop DC gain. All operational amplifier inside the loop filter are ideal have infinite gain to suppress the quantization noise in the signal band of the modulator. The fourth-order noise shaping modulator OBG gain also play an important role, which demands much higher value for 30-bit resolution. The Figure. 3 shows the STF is unity at the low frequencies while the NTF shows a high pass response at OBG value set to 4. The Figure 4 shows the output power spectral density for CIFB topology of the modulator can achieve SNR of 181 dB with an OSR of 131. The Figure 4 shows the OBG value set to 4 for the case of NTF, while the STF have unity gain values according to the simulation environment.

## 1. NON-IDEALITIES SIMULATION

The ideal circuit modeling causes all integrator with very high DC gain while all the feedback path with fixed. To include the circuit non-idealities inside the loop filter as well as low resolution quantizer several non-ideal factor



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included and complete circuit simulated to verify the functionality of the circuit using environment SDToolbox [14]. Which simulates the circuit non-idealities are used. This section will discuss about the circuit non-idealities like thermal noise or kT/C, flicker noise, finite operational amplifier gain, finite slew-rate. The modulator can achieve fourth-order modulator with 4-bit guantizer can achieve signal-to-noise ratio (SNR) of 123 dB with OSR of 32. The proposed modulator with OSR of 64 can achieve SNR of 151 dB. Finally, the modulator can achieve SNR of 182 dB for OSR of 131 with full-scale input of the modulator is 550 mV.

# 4. CONCLUSION

A sixth-order multi-bit delta-sigma modulator modeled, and simulation can achieve higher SNR of 187 dB. The NTF optimization technique employed for maximum performance. resolution. The performance of the operational amplifier investigated for limited DC gain and other circuit non-idealities like thermal noise and flicker noise. Finally, the modulator is simulated for circuit nonidealities of thermal noise and flicker noise.

# 5. ACKNOWLEDGMENT

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# 6. **REFERENCES**

[1] W. Sansen, et al, 1998, Toward sub-1-V analog integrated circuits in submicron standard CMOS technologies, IEEE ISSCC Dig Tech. Papers, pp. 484.

- [2] H. Li, et al, 2011, Novel singleloop multi-bit sigma-delta modulator using OTA sharing technique without DEM, IEICE Electron. Express 8, pp. 2041.
- [3] Bolatkale M, et al,2011, A 4 GHz continuous-time delta-sigma ADC with 70-dB DR and -74 dBFS THD in 125 MHz BW, IEEE Journal of Solid-State Circuits, 46(12), pp. 2857-2868.
- [4] Caldwell T, et al, 2014, A reconfigurable Delta-Sigma ADC with up to 100 MHz using flash reference shuffling, IEEE Trans Circuit & Syst-I: Regular paper, 61(8), pp. 2263-2271.
- [5] Srinivasan V, et al, 2012, A 20 mW 61dB SNDR (60 MHz BW) 1b 3<sup>rd</sup>-order continuous-time deltasigma modulator clocked at 6GHz in 45nm CMOS, IEEE International Solid-State Circuit Conference, pp. 158-160.
- [6] Yoon D, et al, 2015, A continuous-time sturdy-MASH Delta-Sigma Modulator in 28nm CMOS, IEEE Journal of Solid-State Circuit, 50(12), pp. 2880-2890.
- [7] S. Norsworthy, Richard Schreier, G.C. Temes, 1997, Delta-Sigma Data Converter Theory, Design, and Simulation, John Wiley & Sons, Inc., Hoboken, New Jersey.
- [8] Yao Xiao, et al, 2020, A 100-MHz Bandwidth 80-dB Dynamic Range Countinuous-time Delta-Sigma Modulator with a 2.4-GHz Clock Rate, Nanoscale Research Letters.
- [9] Wei Wang, at al., 2019, A 72.6dB SNDR 100MHz BW 16.36mW CTDSM with Preliminary Sampling and Quantization Scheme in Backend Subranging QTZ, IEEE Internation Conference of Solid-State Circuit & Conference (ISSCC), pp. 340-342.



- [10] Sheng-Jui Huang, et al., 2017, A 125MHz-BW 71.9dB-SNDR VCO-Based CT delta-sigma ADC with segmented Phase-Domain ELD Compensation in 16nm, IEEE International Solid-State Circuit Conference (ISSCC), pp. 470-472.
- [11] Yunzhi Dong , et al., 2016, A 930mW 69db-DR 465MHz-BW CT 1-2 MASH ADC in 28 nm CMOS, IEEE International Solid-State Circuit Conference (ISSCC), pp. 278-280.
- [12] **R. Scherier** Delta-Sigma Toolbox ((<u>http://www.mathworks.com/m</u> <u>atlancentral/fileexchange/19-</u> <u>delta-sigma-toolbox</u>).
- [13] Shanti Paven, Richjard Schreier and G.C. Temes, Understanding Delta-Sigma Data Converters Second Edition, IEEE Press Wiley
- [14] S. Brigati SDToolbox (http://www.mathworks.com/ma tlabcentral/fileexchange/2460-sdtoolbox).

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